

Chapt 5 Some Aspects of the Electrodynamics of Continuous Media

§1 Havrilek equations for a dielectric medium

1.1 Electrostatics of dielectrics

Consider the third Havrilek eq. from ch. I, § 3.2:

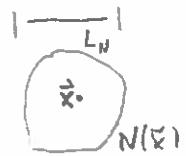
$$\nabla \cdot \vec{E}(\vec{x}) = 4\pi g(\vec{x}),$$

at which a dielectric, i.e., an electric insulator.

Remark: (1) $g(\vec{x})$ is a complicated function that varies rapidly on microscopic length scales.

(2) For macroscopic observations one is not interested in these variations.

microscopic scale def. 1: We define a coarse-grained charge density $\bar{g}(\vec{x})$ by



$$\bar{g}(\vec{x}) = \frac{1}{V_N} \int_{N(\vec{x})}^{} g(\vec{z}) dV_{\vec{z}}$$

where $N(\vec{x})$ is a neighborhood of the point \vec{x} whose volume V_N is large on the microscopic scale, but small on the macroscopic one.

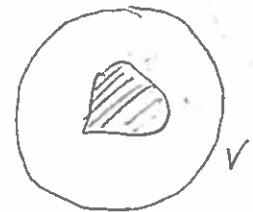
Remark: (3) microscopic scale $\approx \ell$, macroscopic scale $= cm$
 \rightarrow This makes sense.

(4) We also need to coarse-grain the electric field, denote that by \vec{E} as will ($\vec{E} \rightarrow \bar{E}$; call the microscopic field \vec{e} and will $\vec{e} = \vec{E}$)

def. 2: Define the dielectric polarization field $\vec{P}(\vec{x})$ by

$$\vec{P}(\vec{x}) = \begin{cases} \text{the value of } -\nabla \cdot \vec{P}(\vec{x}) = \bar{g}(\vec{x}) \text{ inside the dielectric} \\ 0 \quad \text{outside the dielectric} \end{cases}$$

Remark: (5) Let V be any volume that completely encloses the dielectric body. Then



$$\int_V d\vec{x} \cdot \vec{J}(\vec{x}) = 0$$

$$\text{and } \int_V d\vec{x} \cdot \nabla \cdot \vec{P}(\vec{x}) = \int_V d\vec{\sigma} \cdot \vec{P}(\vec{x}) = 0 \quad (V)$$

def. 3:

$$\vec{D}(\vec{x}) := \vec{E}(\vec{x}) + 4\pi \vec{P}(\vec{x}) \quad \text{is called electric induction}$$

Kern:

The continuum-grained 3rd Maxwell eq. takes the form

$$\nabla \cdot \vec{D}(\vec{x}) = 0 \quad (\delta) \quad \text{proof: } \nabla \cdot \vec{E} = 4\pi \vec{J} = -4\pi \nabla \cdot \vec{P} \rightarrow \nabla \cdot (\vec{E} + 4\pi \vec{P}) = 0$$

Remark: (6) Suppose the dielectric is not uncharged, but carries an "extending" charge density $\rho_{ext}(\vec{x})$ that has been applied by the experimentalist. Then (δ) gets generalized to

$$\nabla \cdot \vec{D}(\vec{x}) = 4\pi \rho_{ext}(\vec{x}) \quad (\delta')$$

(7) Consider the dipole moment of the dielectric:

$$\begin{aligned} d_i &= \int_V d\vec{x} \cdot x_i \vec{J}(\vec{x}) = - \int_V d\vec{x} \cdot x_i \cdot \vec{J} \cdot \vec{P}_i = - \int_V d\vec{x} \cdot \vec{J} \cdot (x_i P_i) + \int_V d\vec{x} \delta_{ij} P_j \\ &= - \underbrace{\int_V d\vec{x} \cdot P_j \cdot x_i}_{(V)} + \int_V d\vec{x} P_i \rightarrow \vec{d} = \int_V d\vec{x} \vec{P}(\vec{x}) \\ &= 0 \text{ since } \vec{P} = 0 \text{ outside the body} \end{aligned}$$

\rightarrow The polarization is the dipole moment density of the dielectric

Comparing this with Maxwell eq. d! § 1.2, yields

$$\nabla \times \vec{E}(\vec{x}) = 0 \quad (\delta \delta)$$

Remark: (8) In order for (δ) and (δδ) to give a complete description, we still need a relation between \vec{D} and \vec{E} .

remark: (9) \vec{P} is the dipole moment density induced by $\vec{E} \rightarrow \vec{P}$ must vanish as $\vec{E} \rightarrow 0$

ansatz: For small \vec{E} , and in a isotropic medium, $\vec{P} \propto \vec{E}$:

$$\boxed{\vec{P}(\vec{x}) = \chi(\vec{x}) \vec{E}(\vec{x})}$$

with χ the dielectric susceptibility.

remark: (10) χ characterizes the medium. In a homogeneous isotropic medium it is a right nr.; in a crystal it is a tensor: $P_i = \chi_{ij} E_j$.

(11) Our ignorance about the microscopic details is hidden in χ .

For the relation between \vec{D} and \vec{E} this applies (in def.)

$$\boxed{\vec{D} = \epsilon \vec{E}}$$

$$\boxed{\epsilon = 1 + 4\pi \chi} \quad \text{the dielectric constant}$$

1.2 Magnetostatics

Now write the first Maxwell eq. from ch 1 § 1.2:

$$\vec{\nabla} \cdot \vec{D}(\vec{x}) = 0$$

Coarse grained cell the coarse-grained magnetic induction
 \vec{H} cell \rightarrow

$$\boxed{\vec{\nabla} \cdot \vec{B}(\vec{x}) = 0}$$

and the fourth Maxwell eq. for static fields:

$$\vec{\nabla} \times \vec{B}(\vec{x}) = \frac{4\pi}{c} \vec{j}(\vec{x})$$

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Statische u. e.
elektrische

with $\tilde{j}(\tilde{x}) = \frac{1}{V_0} \int_{\partial S} d\tilde{s} \tilde{j}(\tilde{s})$ the coarse-grained unit drift.

Remark: (1) In a dielectric, no current can flow

$$\Rightarrow \int_{\partial S} d\tilde{s} \cdot \tilde{j}(\tilde{x}) = 0$$

(v)

$$\int_{\partial S} d\tilde{s} \tilde{\nabla} \cdot \tilde{j}(\tilde{x})$$

v



then the integration is over an arbitrary subvolume of the dielectric $\Rightarrow \tilde{\nabla} \cdot \tilde{j}(\tilde{x}) = 0$

$\Rightarrow \tilde{j}(\tilde{x})$ must be a curl-free field!

def. 1: Define the magnetization field $\tilde{H}(\tilde{x})$ by

$$\tilde{H}(\tilde{x}) = \begin{cases} \text{the solution of } \tilde{\nabla} \times \tilde{H}(\tilde{x}) = \frac{1}{c} \tilde{j}(\tilde{x}) & \text{inside} \\ 0 & \text{outside} \end{cases}$$

def. 2: Define the magnetic field $\tilde{B}(\tilde{x})$ by

$$\tilde{B}(\tilde{x}) := \tilde{D}(\tilde{x}) - 4\pi \tilde{H}(\tilde{x})$$

Under: The coarse-grained force (from e.g. rods)

$$\tilde{\nabla} \times \tilde{H}(\tilde{x}) = 0$$

$$\text{proof: } \tilde{\nabla} \times \tilde{B} = \tilde{\nabla} \times \tilde{D} - 4\pi \tilde{\nabla} \times \tilde{H} = \tilde{\nabla} \times \tilde{D} - \frac{4\pi}{c} \tilde{j} = 0$$

Remark: (2) If we are in draining dielectrics, then we can have no "extended units".

Assumption: For small \tilde{H} , in an isotropic medium,

$$\tilde{H}(\tilde{x}) = \chi_m \tilde{H}(\tilde{x}) \quad \text{with } \chi_m \text{ the } \underline{\text{magnetic}}$$

For the relation between \vec{B} and \vec{H} this implies

$$\vec{B} = \mu_0 \vec{H} + 4\pi \vec{M} = \vec{H} + 4\pi \chi_m \vec{H} \Rightarrow$$

$$\boxed{\vec{B} = \mu \vec{H}}$$

with $\mu = 1 + 4\pi \chi_m$ the magnetic permeability

1.2 Summary of static Maxwell eqs in a dielectric medium

$\oint \oint 1.1, 1.2 \rightarrow$

$$\begin{array}{|c|c|} \hline \nabla \cdot \vec{D}(\vec{x}) = 0 & (1) \\ \hline \nabla \times \vec{E}(\vec{x}) = 0 & (2) \\ \hline \end{array}$$

$$\begin{array}{|c|c|} \hline \nabla \cdot \vec{D}(\vec{x}) = 4\pi \rho_{ext}(\vec{x}) & (3) \\ \hline \nabla \times \vec{B}(\vec{x}) = 0 & (4) \\ \hline \end{array}$$

with \vec{E}, \vec{D} the com-prained electric field and the electric induction, respectively,

and \vec{B}, \vec{H} the com-prained magnetic induction and the magnetic field, respectively

$$\text{and } \boxed{\vec{D} = \epsilon \vec{E}, \quad \vec{B} = \mu \vec{H}} \quad (*)$$

With ED

will ϵ be the dielectric constant and μ the magnetic permeability.

Remark: (1) The invertible relations (*) are valid in a linear media between \vec{D} and \vec{E} , and between \vec{B} and \vec{H} . This is approximately true for small fields. For strong fields (e.g., lasers), ϵ becomes \vec{E} -dependent.

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1.4 Generalization to dynamics: Retarded dielectric response

§ 1.1 \rightarrow For static fields, $\vec{P} = \chi \vec{E}$

remark: (1) In what follows we will ignore the spatial dependence of the fields, as of χ

For a time-dependent field, the polarization \vec{P} cannot follow the field \vec{E} instantaneously \rightarrow The equation for § 1.1 needs to be generalized:

ansatz:

$$\boxed{\vec{P}(t) = \int dt' \Theta(t-t') f(t-t') \vec{E}(t')}$$

remark: (1) Now a function $f(t)$ describes the medium, instead of a single number χ . It is sometimes called a memory function.

(2) We shall soon see that the relation between \vec{P} and \vec{E} is linear

("linear response")

(3) The step function means causality.

(§2) Introduction to the theory of causal functions

2.1 Causal functions (Wk ps 19-21)

Let $f: \mathbb{R} \rightarrow \mathbb{R}$ be a weakly increasing fct. in the sens of 610/[§4.].

Def. 1: $f_{\pm}(t) := \Theta(\pm t) f(t)$ can called the retarded (+) and advanced (-) part of f .

Remark: (1) Thanks 610 \rightarrow Fourier coeffs of f, f_+ , and f_- exist; they are generalized fcts. We denote them by \hat{f}, \hat{f}_{\pm} .

Def. 2: Let $t \in \mathbb{C}$ and define the Laplace transform of $f(t)$, i.e.,

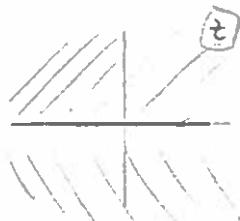
$$F(z) := \pm i \int dt \Theta(\pm t) e^{itz} f(t) \quad \pm \text{ for } \operatorname{Im} z > 0$$

F is called the causal fct. associated with f . t is called complex frequency.

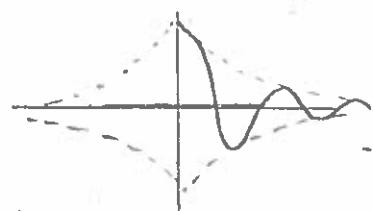
Remark: (2) $t = \omega \pm i\delta$ with $\delta > 0$

$$\rightarrow e^{itz} = e^{i\omega t} e^{\mp \delta t} = e^{i\omega t} e^{-\delta(\pm t)}$$

$\rightarrow F(z)$ is an analytic fct. of t for $\operatorname{Im} t \neq 0$



Example: (1) $f(t) = e^{-i\omega_0 t} e^{-\gamma|t|}$ · damped oscillation



$$\begin{aligned} i \int_0^\infty dt e^{itt - i\omega_0 t - \gamma t} &= \frac{-i}{it - i\omega_0 - \gamma} = \frac{-1}{t - \omega_0 + i\gamma} \\ -i \int_{-\infty}^0 dt e^{itt - i\omega_0 t - \gamma t} &= \frac{-i}{it - i\omega_0 + \gamma} = \frac{-1}{t - \omega_0 - i\gamma} \end{aligned}$$

$$\rightarrow F(z) = \frac{-1}{t - \omega_0 + i\gamma + \gamma^2 \operatorname{Im} z}$$

$F(t)$ is indeed analytic in the upper half plane, as can be analytically continued into the lower half plane when it has a pole at $t = \omega_0 - iy$.

Similarly, f is analytic in the lower half plane and its analytic continuation into the upper half plane has a pole at $t = \omega_0 + iy$.

$F(t)$ consists of two Riemann sheets. It has a branch cut on the real axis, and

$$\underline{F(w \pm i\delta)} = \frac{-1}{w - \omega_0 \pm iy} = \frac{-(w - \omega_0) \pm iy}{(w - \omega_0)^2 + y^2}$$

discontinuity of $\ln F$ on the real axis (if one stays on the sheet

or which the function is analytic in the respective halfplane)

Known: The discontinuity of $F(t)$ across the real axis determines the Fourier transform of $f(t)$:

$$\lim_{\epsilon \rightarrow 0} F(w \pm i\epsilon) = \pm i \hat{f}_{\pm}(w)$$

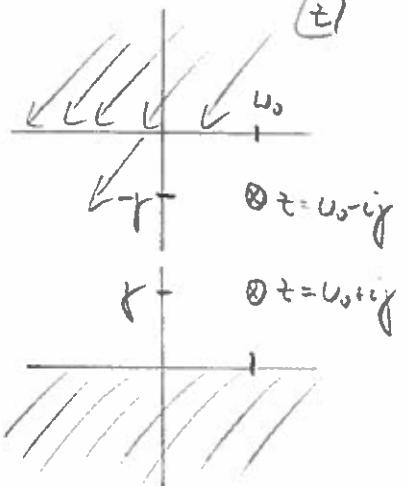
Proof: Let $\Im w + t \geq \epsilon > 0$, i.e., $t = x + iy$ with $x, y \in \mathbb{R}$, $y \geq \epsilon$

$$\rightarrow |e^{i\pi t}| = |e^{ixt} e^{-y\pi t}| \leq e^{-\epsilon \pi t} \quad f(t) \text{ decays}$$

$$\rightarrow |F(t)| = \left| i \int_0^\infty dt e^{i\pi t} f(t) \right| \leq \int_0^\infty dt e^{-\epsilon \pi t} f(t) < \infty$$

$\rightarrow F(w \pm i\epsilon)$ is bounded and analytic

$\rightarrow F(w \pm i\epsilon)$ is a regular generalized fct. in the sense of PWL § 610 uZ § 4.4



Now let $\hat{g}(w)$ be a test fct., $g \in \mathcal{F}$, with nn of 610/§4
ad wieder $\begin{array}{l} \text{Period} = \frac{\pi}{\omega} \\ 610 \text{ u. } 2 \text{ §4.3} \end{array}$

$$\int dw F(w+i\varepsilon) \hat{g}^*(w) \stackrel{\varepsilon \rightarrow 0}{\longrightarrow} 2\pi i \int_0^\infty dt e^{-\varepsilon t} f(t) g^*(-t)$$

$$\stackrel{\varepsilon \rightarrow 0}{\longrightarrow} 2\pi i \int_0^\infty dt f(t) g^*(-t)$$

$$\text{Int } |f(t) e^{-\varepsilon t} g^*(-t)| \leq |f(t)| \cdot |g^*(-t)|$$

\rightarrow The integral with $\varepsilon \rightarrow 0$ $\rightarrow \lim_{\varepsilon \rightarrow 0} \int_0^\infty dt$ commutes

$$\rightarrow \lim_{\varepsilon \rightarrow 0} \int dw F(w+i\varepsilon) \hat{g}^*(w) = i \int dw \hat{f}_+(w) \hat{g}^*(w) + g \in \mathcal{F}$$

$$\rightarrow \underbrace{\lim_{\varepsilon \rightarrow 0} F(w+i\varepsilon)}_{= \hat{f}_+(w)} = \hat{f}_+(w) \text{ with nn of 610 u. 2 §4.4}$$

Analogous proof for $\varepsilon < 0$.

wollen:

$$\hat{f}(w) = \hat{f}_+(w) + \hat{f}_-(w) = -i [F(w+i0) - F(w-i0)]$$

$$\text{ad } \hat{f}_+(w) - \hat{f}_-(w) = -i [F(w+i0) + F(w-i0)]$$

example: (2) Winkler example (1) with $f \rightarrow 0$. $\rightarrow f(t) = e^{-i\omega_0 t}$

$$\rightarrow \hat{f}(w) = 2\pi \delta(w-\omega_0) \text{ ad } F(t) = \frac{-1}{t-\omega_0}$$

$$\rightarrow -i [F(w+i0) - F(w-i0)] = \lim_{\varepsilon \rightarrow 0} i \left(\frac{-1}{w-\omega_0+i\varepsilon} + \frac{1}{w-\omega_0-i\varepsilon} \right)$$

$$= \lim_{\varepsilon \rightarrow 0} \frac{-2i\varepsilon}{i} \frac{-1}{(\omega-\omega_0)^2+\varepsilon^2} = \lim_{\varepsilon \rightarrow 0} \frac{2\varepsilon}{(\omega-\omega_0)^2+\varepsilon^2} = \underline{2\pi \delta(\omega-\omega_0)}$$

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2.2 Spectrum and reactive part of causal functions

def. 1: Let $F(t)$ be a causal function. We define

$$\boxed{F''(w) := \frac{1}{2i} [F(w+i0) - F(w-i0)]}$$

$$\boxed{F'(w) := \frac{1}{2} [F(w+i0) + F(w-i0)]}$$

F'' is called spectrum, or spectral part, or disipative part
 F' is called reactive part of $F(t)$.

remark: (1) Our notation implies that $f(t), \hat{f}(w), F(z), F''(w), F'$ are different fcts. One often writes $f(t), f(w), f(z), f''(w), f'(w)$ to distinguish the fcts only by their arguments and the prime and double prime superscripts.

(2) $F''(w)$ is given by the discontinuity of $F(z)$ across the real axis; $F'(w)$ by the average of $F(z)$ across the real axis.

$$(3) \boxed{F(w \pm i0) = F'(w) \pm iF''(w)}$$

$$(4) \boxed{\begin{aligned} \hat{f}(w) &= \hat{f}_+(w) + \hat{f}_-(w) = \mathcal{Z}F''(w) \\ \tilde{f}(w) &= \hat{f}_+(w) - \hat{f}_-(w) = \mathcal{Z}F'(w) \end{aligned}}$$

(-i) ??

Week 9

about 18 (§ 49, 50, 51)

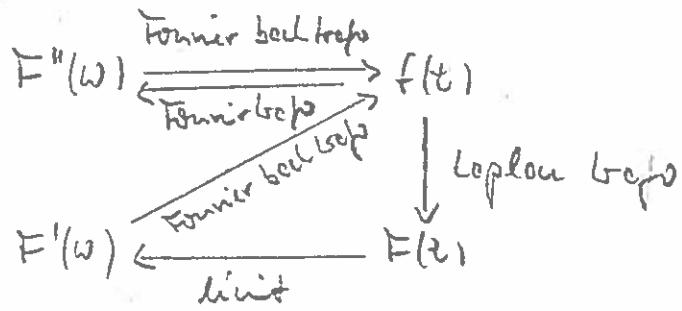
(5) In general, $F'(w)$ and $F''(w)$ are 0-valued fcts.

theorem: (a) The spectrum $F''(w)$ uniquely determines $f(t), F(z)$ and $F'(w)$

(b) $F'(w)$ uniquely determines $F(z)$ and $F''(w)$.

remark: (6) This is an extremely important result, even though it follows immediately from the definition

proof:



Remark: (7) We need the entire fct. $F''(w)$, i.e., $F''(w) \neq 0$, in order to determine $F'(w)$ for a given w .

Woolley: For a given generalized fct. $F''(w)$, there exists at most one fct. $F(z)$ such that

- (i) $F(z)$ is analytic for $\operatorname{Im} z \neq 0$ and
- (ii) $F(z) \rightarrow 0$ for $\operatorname{Im} z \rightarrow \infty$.

proof: Let F_1, F_2 be two such fcts and consider $G = F_1 - F_2$

$\Rightarrow G(z)$ is analytic for $\operatorname{Im} z \neq 0$ and $G(z) \rightarrow 0$ for $\operatorname{Im} z \rightarrow \infty$

$$\begin{aligned} \text{But } G(w+i0) - G(w-i0) &= F_1(w+i0) - F_2(w+i0) \\ &\quad - F_1(w-i0) + F_2(w-i0) \\ &= 2i [F_1''(w) - F_2''(w)] = 2i [F''(w) - F''(w)] = 0 \end{aligned}$$

$\Rightarrow G(z)$ is analytic $\forall z$ (the rest is left to the reader!)

$\rightsquigarrow G(z)$ is a polynomial in z

But $G(z) \rightarrow 0$ for $\operatorname{Im} z \rightarrow \infty \Rightarrow G(z) \equiv 0$

A know in complex analysis

2.3 Hilbert-Stieltjes transformation

def. Let $f: \mathbb{R} \rightarrow \mathbb{C}$ be a measurable fct. and let

$$\boxed{F(z) = \int_{\mathbb{R}} \frac{f(u)}{u-z} du} \quad \text{exists for } \operatorname{Im} z \neq 0$$

then $F(z)$ is called the Hilbert-Stieltjes transform of $f(u)$.

example: (1) $f(u) = \delta(u-u_0) \Rightarrow F(z) = \frac{1}{u_0 - z}$

remark: (1) $f \rightarrow F$ is a integral transf. that maps $f: \mathbb{R} \rightarrow \mathbb{C}$ onto $F: \mathbb{C} \rightarrow \mathbb{C}$.

(2) $F(z)$ is analytic for $\operatorname{Im} z \neq 0$ and has a branch cut for $\operatorname{Im} z = 0$.

proposition 1: (a) If $f(u)$ is even (odd), then $F(z)$ is odd (even).

(b) If $f(u) \rightarrow F(z)$, then $f^*(u) \rightarrow F(z^*)^*$

(c) If $f(u) \in \mathbb{R}$, then $F(z) = F(z^*)^*$

If $f(u) \in i\mathbb{R}$, then $F(z) = -F(z^*)^*$

proof: (a) Let $f(u) = f(-u) \Rightarrow F(-z) = \int_{\mathbb{R}} \frac{du}{u+z} \cdot \int_{\mathbb{R}} \frac{f(u)}{u-z} du = \int_{\mathbb{R}} \frac{du}{u-z} \cdot f(-u) = -F(z)$

$$(b) \int_{\mathbb{R}} \frac{du}{u-z} \frac{f(u)^*}{u-z} = \left(\int_{\mathbb{R}} \frac{du}{u-z} \frac{f(u)}{u-z^*} \right)^* = F(z^*)^*$$

$$(c) \text{Let } f(u) \in \mathbb{R} \Rightarrow F(z^*) = \int_{\mathbb{R}} \frac{du}{u-z^*} \frac{f(u)}{u-z} = \left(\int_{\mathbb{R}} \frac{du}{u-z} \frac{f(u)}{u-z} \right)^* \cdot F(z)$$

$$\text{Let } f(u) \in i\mathbb{R} \Rightarrow F(z^*) = \int_{\mathbb{R}} \frac{du}{u-z^*} \frac{-f(u)^*}{u-z} \cdot -F(z)^*$$

Lemma 1: (Sokhotski-Plemelj)

If $F(z)$ is bounded for $|Im z| \geq \varepsilon$, then

$$\lim_{\varepsilon \rightarrow 0} \frac{1}{2i} [F(u+i\varepsilon) - F(u-i\varepsilon)] = f(u)$$

Simplified proof: Let $g(u)$ be a test fct. Then

$$\begin{aligned} \frac{1}{2i} \int du g(u) [F(u+i\varepsilon) - F(u-i\varepsilon)] &= \frac{1}{2i} \int du g(u) \int_{\mathbb{R}} \frac{dx}{x-u-i\varepsilon} f(x) \left(\frac{1}{x-u-i\varepsilon} - \frac{1}{x-u+i\varepsilon} \right) \\ &\cdot \frac{1}{2i} \int du g(u) \int_{\mathbb{R}} \frac{dx}{x-u-i\varepsilon} f(x) \frac{i\varepsilon}{(x-u)^2 + \varepsilon^2} \xrightarrow{\varepsilon \rightarrow 0} \int du g(u) \int_{\mathbb{R}} \frac{dx}{x-u} f(x) \delta(x-u) \\ &= \int du g(u) f(u) \quad \# \end{aligned}$$

Lemma 2: Let $F: \mathbb{C} \rightarrow \mathbb{C}$ be a fct with the properties

(1) $F(z)$ is analytic for $|m z| > 0$

(2) $F(z) \rightarrow 0$ for $|z| \rightarrow \infty$

(3) $\lim_{\varepsilon \rightarrow 0} \frac{1}{2i} [F(u+i\varepsilon) - F(u-i\varepsilon)]$ exists and defines a generalized fct. $f(u)$.

Then $F(z)$ can be written in H-S form, i.e., $F(z) = \int_0^\infty \frac{f(u)}{u-z} du$ and $F(z)$ is the unique H-S repr. of $f(u)$.

Proof: books

Proposition 2: Let $F(z)$ be the H-S repr. of a real generalized fct. $f(u) \neq 0$. Then

$$f(u) \geq 0 \text{ iff } (\operatorname{Im} F(z) \geq 0 \text{ for } \operatorname{Im} z \geq 0)$$

$$\begin{aligned}
 \text{Proof: } F(x+iy) &= \int_{\sigma} \frac{f(u)}{u-(x+iy)} du = \int_{\sigma} \frac{(u-x+iy)f(u)}{(u-x)^2+y^2} du \\
 &= \underbrace{\int_{\sigma} \frac{(u-x)f(u)}{(u-x)^2+y^2} du}_{\in \mathbb{R}} + iy \underbrace{\int_{\sigma} \frac{f(u)}{(u-x)^2+y^2} du}_{>0 \text{ for } f>0}
 \end{aligned}$$

Therefore, $f(u) \geq 0$ implies $\operatorname{Im}\left(\frac{1}{2}F(x+iy)\right) \geq 0$

and $f(z) \neq 0$ implies $\operatorname{Im}\left(\frac{1}{2}F(x+iy)\right) > 0$

But $y = m + t \Rightarrow f(u) \geq 0$ implies $\operatorname{Im} F(z) \geq 0$ for $m+t \geq 0$

Now let $\operatorname{Im} F(z) \geq 0$ for $m+t \geq 0$

$$\operatorname{Im} \frac{f(u)}{z-i} = \frac{1}{2i} [F(u+i0) - F(u-i0)] = \frac{1}{2} [\operatorname{Im} F(u+i0) - \operatorname{Im} F(u-i0)] \geq 0$$

def. 2: The H-S graphs of non-negative functions are called positive functions.

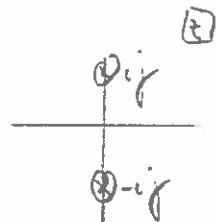
Remark: (i) Positivity in this sense is equivalent to $\operatorname{Im} F(z) \geq 0$ for $m+t \geq 0$.

2.4 Some examples of Hilbert-Schmidt's graphs

(1) Hyperbola for $m+t \geq 0$

$$F(z) = \frac{-1}{z+iy} \quad \text{is analytic for } m+t \neq 0$$

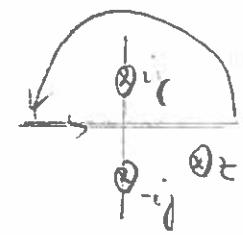
$\text{and } \rightarrow 0 \text{ for } |z| \rightarrow \infty$



$$\text{and } \frac{1}{2i} [F(u+i0) - F(u-i0)] = \frac{1}{2i} \left(\frac{-1}{u+iy} + \frac{1}{u-iy} \right) = \frac{2}{u^2+y^2} \text{ which}$$

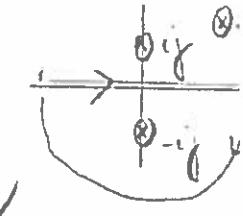
$$\rightarrow \frac{-1}{z+iy} = \int_{\sigma} \frac{du}{\sigma} \frac{1}{u-z} \frac{2}{u^2+y^2}$$

Durch: $\int \frac{du}{\sigma} \frac{1}{w-t} \frac{\kappa}{u^2 + \gamma^2} = \int \frac{du}{\sigma} \frac{1}{w-t} \frac{\kappa}{(u+i\gamma)(u-i\gamma)}$



$$= \oint \frac{du}{\sigma} \frac{1}{u-i\gamma} \frac{1}{u-t} \cdot \frac{-1}{t-i\gamma} \quad \text{for } \Im t < 0$$

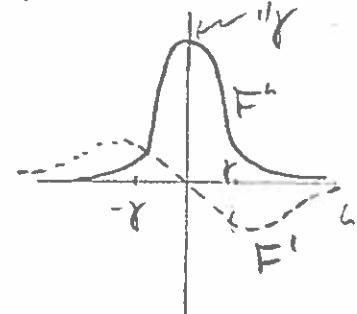
$$\text{or } \oint \frac{du}{\sigma} \frac{1}{u+i\gamma} \frac{1}{u-t} = \frac{-1}{t+i\gamma} \quad \text{for } \Im t > 0$$



Remark: (1) $\oint du/\sigma \rightarrow$ The spectrum of the reaction part of $F(t)$ in

$$\underline{F''(\omega)} = \frac{\kappa}{\underline{u^2 + \gamma^2}}$$

$$\underline{F'(w)} = \frac{1}{2} \left(\frac{-1}{w+i\gamma} - \frac{1}{w-i\gamma} \right) = \frac{-w}{\underline{w^2 + \gamma^2}}$$



2.5 Spectral representation, and Kramers-Kronig relations

Combining the results of the previous paragraphs, we have the following

Kronecker: A causal fct. $F(t)$ can be written in terms of its spectra $F''(\omega)$

$$\boxed{F(t) = \int \frac{du}{\sigma} \frac{F''(u)}{u-t}} \quad (\star)$$

Remark: (1) (\star) is called spectral representation or Lehmann representation of $F(t)$.

(2) For $t = \omega' \pm i\varepsilon$ the denominator takes the form

$$\frac{1}{w-\omega' \pm i\varepsilon} = \frac{w-\omega' \pm i\varepsilon}{(w-\omega')^2 + \varepsilon^2} = \frac{w-\omega'}{(w-\omega')^2 + \varepsilon^2} \pm i \frac{\varepsilon}{(w-\omega')^2 + \varepsilon^2}$$

(3) We know that $\lim_{\varepsilon \rightarrow 0^+} \frac{\varepsilon}{(w-\omega')^2 + \varepsilon^2} = \delta(\omega - \omega')$ is a fundamental

Def. I: $\lim_{\varepsilon \rightarrow 0} \frac{x}{x^2 + \varepsilon^2}$ is called the principal-value generalized fct

$$\text{et. or with } \int dx f(x) \lim_{\varepsilon \rightarrow 0} \frac{x}{x^2 + \varepsilon^2} =: \oint dx \frac{f(x)}{x}$$

Remark: (4) One can show that this exists for large classes of fcts. $f(x)$, and one often writes $\int dx$ instead of $\oint dx$.

Woolley: The spectrum of the nuclear part of the causal fct. $F(z)$ is related by the Kramers-Kronig relations

$$\boxed{F'(u) = \int \frac{dx}{\pi} \frac{F^*(x)}{x-u}, \quad F^*(u) = - \int \frac{dx}{\pi} \frac{F'(x)}{x-u}}$$

Proof:

$$\begin{aligned} F'(u) &= \frac{1}{2} [F(u+i0) + F(u-i0)] = \frac{1}{2} \int \frac{dx}{\pi} F^*(x) \left(\frac{1}{x-u-i0} + \frac{1}{x-u+i0} \right) \\ &\Rightarrow \frac{1}{2} \int \frac{dx}{\pi} F^*(x) \frac{2(x-u)}{(x-u)^2 + 0^2} = \oint \frac{dx}{\pi} \frac{F^*(x)}{x-u} \end{aligned}$$

Now consider $\tilde{F}(t) = \operatorname{sp}(\operatorname{ant}) F(t)$. The premises of [2.3] theorem 2 are fulfilled $\Rightarrow \tilde{F}(t)$ can be written in LHS form, and $\tilde{F}(u \pm i0) = \pm F(u \pm i0) \stackrel{H2}{=} \pm F'(u) + iF^*(u)$

$$\Rightarrow \tilde{F}^*(u) = \frac{1}{2i} [\tilde{F}(u+i0) - \tilde{F}(u-i0)] = \frac{1}{2i} F'(u)$$

$$\Leftrightarrow \tilde{F}'(u) = \frac{1}{2} [\tilde{F}(u+i0) + \tilde{F}(u-i0)] = iF^*(u)$$

$$\Rightarrow \underline{\underline{F^*(u)}} = -i \tilde{F}'(u) = -i \int \frac{dx}{\pi} \frac{\tilde{F}^*(x)}{x-u} = - \int \frac{dx}{\pi} \frac{F'(x)}{x-u}$$

2.6 Applicatie: The dielectric function

§1.4 \rightarrow The linear relation between the polarization \vec{P} and the electric field \vec{E} is given by a fct. $f_+(t) \rightarrow$ All results of §2 apply!

$$\text{Lepton loops} \rightarrow \vec{P}(t) = X(t) \vec{E}(t)$$

will $X(t)$ a causal fct.

$$\rightarrow X(\omega \pm i0) = X'(\omega) \pm iX''(\omega) \quad \text{when } X', X'' \text{ obey Kramers-Kronig}$$

$$\S1.1 \rightarrow \epsilon(t) = 1 + 4\pi X(t)$$

$$\rightarrow \epsilon'(\omega) = 1 + 4\pi X'(\omega), \quad \epsilon''(\omega) = 4\pi X''(\omega)$$

$$\text{KK} \rightarrow \boxed{\epsilon'(\omega) = 1 + 4\pi X'(\omega) = 1 + 4\pi \int_{-\infty}^{\omega} \frac{dx}{x} \frac{X''(x)}{x-\omega}} = \boxed{1 + \int_{-\infty}^{\omega} \frac{dx}{x} \frac{\epsilon''(x)}{x-\omega}}$$

$$\boxed{\epsilon''(\omega) = 4\pi X''(\omega) = -4\pi \int_{-\infty}^{\omega} \frac{dx}{x} \frac{X'(x)}{x-\omega}} = \boxed{\int_{-\infty}^{\omega} \frac{dx}{x} \frac{1-\epsilon'(x)}{x-\omega}} \quad \begin{matrix} \text{Kronig 1926} \\ \text{Kramers 1928} \end{matrix}$$

Remark: (1) by using consideration $\rightarrow \epsilon'(\omega)$ red and even
 $\epsilon''(\omega)$ red and odd

Example: Measured $\epsilon(\omega)$ of SiN films

↑ This is wrong!
 Check how $X(t)$ is
 defined in this worksheet!
 Is this in the Kramers
 loops using a different
 definition??

Infrared dielectric properties of low-stress silicon nitride

Giuseppe Cataldo,^{1,2,*} James A. Beall,³ Hsiao-Mei Cho,³ Brendan McAndrew,¹ Michael D. Niemack,³ and Edward J. Wollack¹

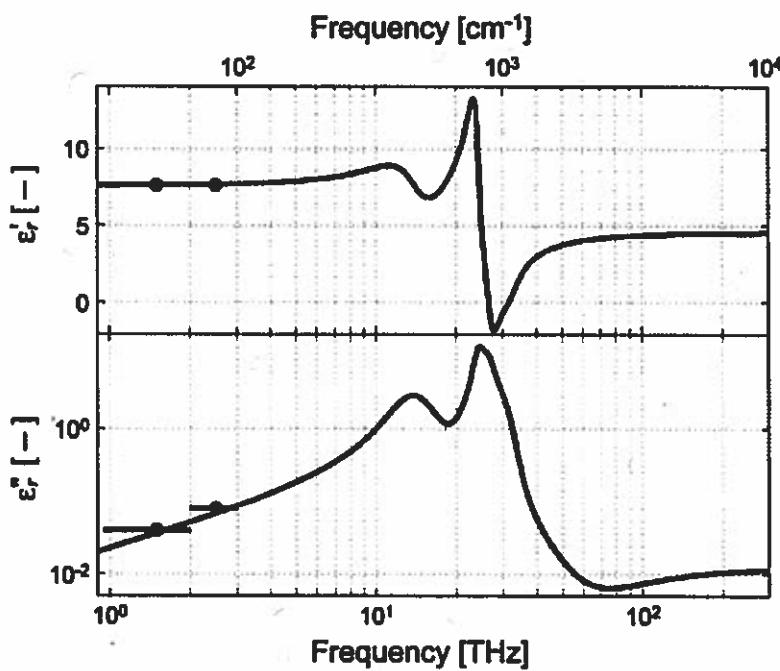


Fig. 3. (Color online) Real and imaginary parts (solid red curves) of the dielectric function as extracted from the data